

# 50 mA, 100 mA and 150 mA CMOS LDOs with Shutdown and ERROR Output

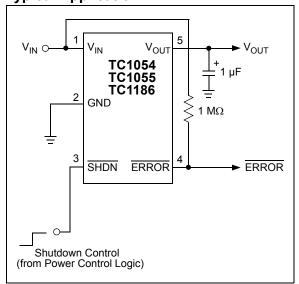
#### **Features**

- · Low Ground Current for Longer Battery Life
- · Low Dropout Voltage
- Choice of 50 mA (TC1054), 100 mA (TC1055) and 150 mA (TC1186) Output
- · High Output Voltage Accuracy
- · Standard or Custom Output Voltages:
  - 1.8V, 2.5V, 2.6V, 2.7V, 2.8V, 2.85V, 3.0V, 3.3V, 3.6V, 4.0V, 5.0V
- · Power-Saving Shutdown Mode
- ERROR Output Can Be Used as a Low Battery Detector or Microcontroller Reset Generator
- · Overcurrent and Overtemperature Protection
- 5-Pin SOT-23 Package
- · Pin Compatible Upgrades for Bipolar Regulators

#### **Applications**

- · Battery Operated Systems
- · Portable Computers
- · Medical Instruments
- Instrumentation
- · Cellular/GSM/PHS Phones
- · Linear Post-Regulators for SMPS
- Pagers

#### Typical Application



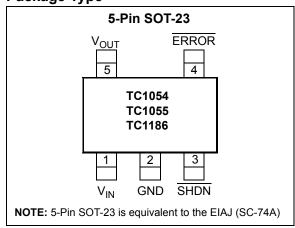
#### **General Description**

The TC1054, TC1055 and TC1186 are high accuracy (typically  $\pm 0.5\%$ ) CMOS upgrades for older (bipolar) low dropout regulators. Designed specifically for battery-operated systems, the devices' CMOS construction minimizes ground current, extending battery life. Total supply current is typically 50  $\mu$ A at full load (20 to 60 times lower than in bipolar regulators).

The devices' key features include low noise operation, low dropout voltage — typically 85 mV (TC1054), 180 mV (TC1055) and 270 mV (TC1186) at full load — and fast response to step changes in load. An error output (ERROR) is asserted when the devices are out-of-regulation (due to a low input voltage or excessive output current). ERROR can be used as a low battery warning or as a processor RESET signal (with the addition of an external RC network). Supply current is reduced to 0.5  $\mu A$  (max), with both  $V_{OUT}$  and ERROR disabled when the shutdown input is low. The devices incorporate both over-temperature and over-current protection.

The TC1054, TC1055 and TC1186 are stable with an output capacitor of only 1  $\mu$ F and have a maximum output current of 50 mA, 100 mA and 150 mA, respectively. For higher output current regulators, please refer to the TC1173 ( $I_{OUT}$  = 300 mA) data sheet (DS21632).

#### Package Type



# 1.0 ELECTRICAL CHARACTERISTICS

#### **Absolute Maximum Ratings †**

Input Voltage	6.5V
Output Voltage	(-0.3V) to (V <sub>IN</sub> + 0.3V)
Power Dissipation	Internally Limited (Note 6)
Maximum Voltage on Any Pin	V <sub>IN</sub> +0.3V to -0.3V
Operating Junction Temperature F	Range40°C < T <sub>J</sub> < 125°C
Storage Temperature	65°C to +150°C

† Stresses above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only and functional operation of the device at these or any other conditions above those indicated in the operation sections of the specifications is not implied. Exposure to Absolute Maximum Rating conditions for extended periods may affect device reliability.

#### DC CHARACTERISTICS

Electrical Specifications: Unless otherwise noted,  $V_{IN} = V_{OUT} + 1V$ ,  $I_L = 100 \mu A$ ,  $C_L = 3.3 \mu F$ ,  $\overline{SHDN} > V_{IH}$ ,  $T_A = +25^{\circ}C$ .

<b>Boldface</b> type specifications apply for junction temperatures of -40°C to +125°C.						
Parameters	Sym	Min	Тур	Max	Units	Conditions
Input Operating Voltage	V <sub>IN</sub>	2.7	_	6.0	V	Note 8
Maximum Output Current	I <sub>OUTMAX</sub>	50	_	_	mA	TC1054
	33111111	100	_	_		TC1055
		150	_	_		TC1186
Output Voltage	$V_{OUT}$	V <sub>R</sub> - 2.5%	$V_R \pm 0.5\%$	V <sub>R</sub> + 2.5%	V	Note 1
V <sub>OUT</sub> Temperature Coefficient	TCV <sub>OUT</sub>	_	20	_	ppm/°C	Note 2
			40			
Line Regulation	$\Delta V_{OUT}/\Delta V_{IN}$	_	0.05	0.35	%	$(V_R + 1V) \le V_{IN} \le 6V$
Load Regulation:	$\Delta V_{OUT}/V_{OUT}$					(Note 3)
TC1054; TC1055		_	0.5	2	%	$I_L = 0.1 \text{ mA to } I_{OUTMAX}$
TC1186			0.5	3		$I_L = 0.1 \text{ mA to } I_{OUTMAX}$
Dropout Voltage:	$V_{IN}$ - $V_{OUT}$	_	2	_	mV	I <sub>L</sub> = 100 μA
		_	65	_		$I_L = 20 \text{ mA}$
		_	85	120		$I_L = 50 \text{ mA}$
TC1055; TC1186		_	180	250		I <sub>L</sub> = 100 mA
TC1186		_	270	400		I <sub>L</sub> = 150 mA ( <b>Note 4</b> )
Supply Current	I <sub>IN</sub>	_	50	80	μA	$\overline{SHDN} = V_{IH}$ , $I_L = 0 \mu A$ (Note 9)
Shutdown Supply Current	I <sub>INSD</sub>	_	0.05	0.5	μΑ	SHDN = 0V
Power Supply Rejection Ratio	PSRR	1	64	1	dB	$f \le 1 \text{ kHz}$
Output Short Circuit Current	I <sub>OUTsc</sub>	_	300	450	mA	V <sub>OUT</sub> = 0V
Thermal Regulation	$\Delta V_{OUT}/\Delta P_{D}$	_	0.04	_	V/W	Notes 5, 6
Thermal Shutdown Die Temperature	T <sub>SD</sub>	_	160	_	°C	
Thermal Shutdown Hysteresis	$\DeltaT_{SD}$	1	10	_	°C	

Note 1: V<sub>R</sub> is the regulator output voltage setting. For example: V<sub>R</sub> = 1.8V, 2.5V, 2.7V, 2.85V, 3.0V, 3.3V, 3.6V, 4.0V, 5.0V.

- 2:  $TC V_{OUT} = \frac{(V_{OUTMAX} V_{OUTMIN})x \cdot 10^6}{V_{OUT} x \Delta T}$
- 3: Regulation is measured at a constant junction temperature using low duty cycle pulse testing. Load regulation is tested over a load range from 0.1 mA to the maximum specified output current. Changes in output voltage due to heating effects are covered by the thermal regulation specification.
- 4: Dropout voltage is defined as the input to output differential at which the output voltage drops 2% below its nominal value.
- 5: Thermal Regulation is defined as the change in output voltage at a time T after a change in power dissipation is applied, excluding load or line regulation effects. Specifications are for a current pulse equal to I<sub>LMAX</sub> at V<sub>IN</sub> = 6V for T = 10 ms.
- **6:** The maximum allowable power dissipation is a function of ambient temperature, the maximum allowable junction temperature and the thermal resistance from junction-to-air (i.e.,  $T_A$ ,  $T_J$ ,  $\theta_{JA}$ ). Exceeding the maximum allowable power dissipation causes the device to initiate thermal shutdown. Please see **Section 5.0 "Thermal Considerations"**, "Thermal Considerations", for more details.
- 7: Hysteresis voltage is referenced by V<sub>R</sub>.
- 8: The minimum  $V_{IN}$  has to justify the conditions:  $V_{IN} \ge V_R + V_{DROPOUT}$  and  $V_{IN} \ge 2.7V$  for  $I_L = 0.1$  mA to  $I_{OUT_{MAX}}$ .
- 9: Apply for junction temperatures of -40C to +85C.

#### DC CHARACTERISTICS (CONTINUED)

**Electrical Specifications:** Unless otherwise noted,  $V_{IN} = V_{OUT} + 1V$ ,  $I_L = 100 \mu A$ ,  $C_L = 3.3 \mu F$ ,  $\overline{SHDN} > V_{IH}$ ,  $T_A = +25^{\circ}C$ . **Boldface** type specifications apply for junction temperatures of -40°C to +125°C.

31 1 113 3 1						
Parameters	Sym	Min	Тур	Max	Units	Conditions
Output Noise	eN	_	260	_	nV/√Hz	I <sub>L</sub> = I <sub>OUTMAX</sub>
SHDN Input						
SHDN Input High Threshold	V <sub>IH</sub>	45	_	1	%V <sub>IN</sub>	V <sub>IN</sub> = 2.5V to 6.5V
SHDN Input Low Threshold	$V_{IL}$	_	_	15	%V <sub>IN</sub>	V <sub>IN</sub> = 2.5V to 6.5V
ERROR Output						
Minimum Vin Operating Voltage	$V_{INMIN}$	1.0	_	1	V	
Output Logic Low Voltage	V <sub>OL</sub>	_	_	400	mV	1 mA Flows to ERROR
ERROR Threshold Voltage	$V_{TH}$	_	0.95 x V <sub>R</sub>	_	V	See Figure 4-2
ERROR Positive Hysteresis	V <sub>HYS</sub>	_	50	_	mV	Note 7
V <sub>OUT</sub> to ERROR Delay	t <sub>DELAY</sub>	_	2.5	_	ms	$V_{OUT}$ falling from $V_R$ to $V_R$ - 10%

Note 1:  $V_R$  is the regulator output voltage setting. For example:  $V_R$  = 1.8V, 2.5V, 2.7V, 2.85V, 3.0V, 3.3V, 3.6V, 4.0V, 5.0V.

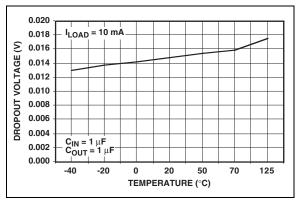
2: TC V<sub>OUT</sub> = 
$$\frac{(V_{OUTMAX} - V_{OUTMIN})x \cdot 10^6}{V_{OUT} \times \Delta T}$$

- 3: Regulation is measured at a constant junction temperature using low duty cycle pulse testing. Load regulation is tested over a load range from 0.1 mA to the maximum specified output current. Changes in output voltage due to heating effects are covered by the thermal regulation specification.
- 4: Dropout voltage is defined as the input to output differential at which the output voltage drops 2% below its nominal value
- 5: Thermal Regulation is defined as the change in output voltage at a time T after a change in power dissipation is applied, excluding load or line regulation effects. Specifications are for a current pulse equal to I<sub>LMAX</sub> at V<sub>IN</sub> = 6V for T = 10 ms.
- **6:** The maximum allowable power dissipation is a function of ambient temperature, the maximum allowable junction temperature and the thermal resistance from junction-to-air (i.e.,  $T_A$ ,  $T_J$ ,  $\theta_{JA}$ ). Exceeding the maximum allowable power dissipation causes the device to initiate thermal shutdown. Please see **Section 5.0 "Thermal Considerations"**, "Thermal Considerations", for more details.
- 7: Hysteresis voltage is referenced by V<sub>R</sub>.
- 8: The minimum  $V_{IN}$  has to justify the conditions:  $V_{IN} \ge V_R + V_{DROPOUT}$  and  $V_{IN} \ge 2.7V$  for  $I_L = 0.1$  mA to  $I_{OUT_{MAX}}$ .
- **9:** Apply for junction temperatures of -40C to +85C.

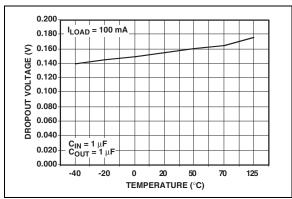
#### 2.0 TYPICAL PERFORMANCE CURVES

**Note:** The graphs and tables provided following this note are a statistical summary based on a limited number of samples and are provided for informational purposes only. The performance characteristics listed herein are not tested or guaranteed. In some graphs or tables, the data presented may be outside the specified operating range (e.g., outside specified power supply range) and therefore outside the warranted range.

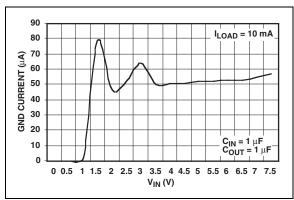
Note: Unless otherwise indicated,  $V_{IN} = V_{OUT} + 1V$ ,  $I_L = 100 \mu A$ ,  $C_L = 3.3 \mu F$ ,  $\overline{SHDN} > V_{IH}$ ,  $T_A = +25^{\circ}C$ .



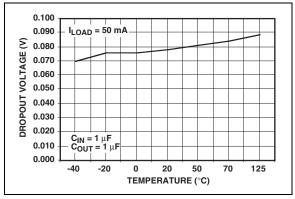
**FIGURE 2-1:** Dropout Voltage vs. Temperature ( $I_{LOAD} = 10 \text{ mA}$ ).



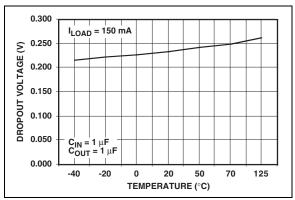
**FIGURE 2-2:** Dropout Voltage vs. Temperature ( $I_{LOAD} = 100 \text{ mA}$ ).



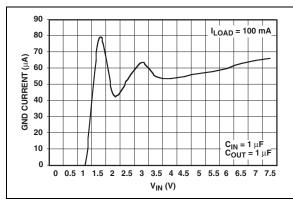
**FIGURE 2-3:** Ground Current vs.  $V_{IN}$  ( $I_{LOAD} = 10 \text{ mA}$ ).



**FIGURE 2-4:** Dropout Voltage vs. Temperature ( $I_{LOAD} = 50 \text{ mA}$ ).

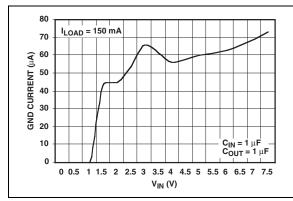


**FIGURE 2-5:** Dropout Voltage vs. Temperature ( $I_{LOAD} = 150 \text{ mA}$ ).



**FIGURE 2-6:** Ground Current vs.  $V_{IN}$  ( $I_{LOAD} = 100 \text{ mA}$ ).

**Note:** Unless otherwise indicated,  $V_{IN} = V_{OUT} + 1V$ ,  $I_L = 100 \mu A$ ,  $C_L = 3.3 \mu F$ ,  $\overline{SHDN} > V_{IH}$ ,  $T_A = +25 ^{\circ}C$ .



**FIGURE 2-7:** Ground Current vs.  $V_{IN}$  ( $I_{LOAD} = 150 \text{ mA}$ ).

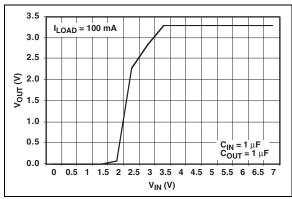


FIGURE 2-8:  $V_{OUT}$  vs.  $V_{IN}$  ( $I_{LOAD}$  = 100 mA).

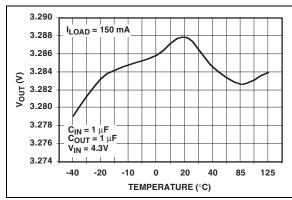
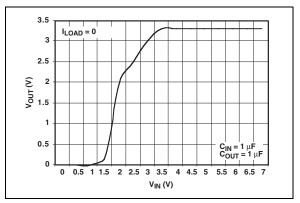
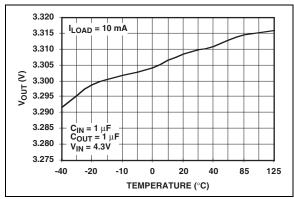


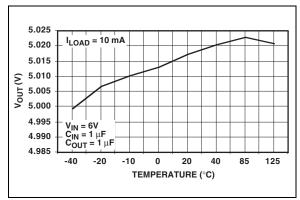
FIGURE 2-9:  $V_{OUT}$  vs.  $V_{IN}$  ( $I_{LOAD}$  = 150 mA).



**FIGURE 2-10:**  $V_{OUT}$  vs.  $V_{IN}$   $(I_{LOAD} = 0 \text{ mA}).$ 

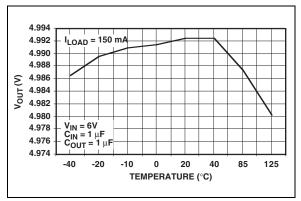


**FIGURE 2-11:** Output Voltage (3.3V) vs. Temperature ( $I_{LOAD} = 10 \text{ mA}$ ).

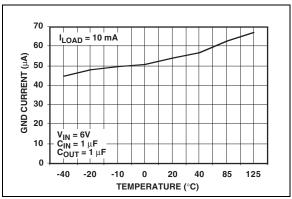


**FIGURE 2-12:** Output Voltage (5V) vs. Temperature ( $I_{LOAD}$  = 10 mA).

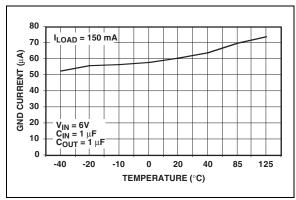
**Note:** Unless otherwise indicated,  $V_{IN} = V_{OUT} + 1V$ ,  $I_L = 100 \mu A$ ,  $C_L = 3.3 \mu F$ ,  $\overline{SHDN} > V_{IH}$ ,  $T_A = +25^{\circ}C$ .



**FIGURE 2-13:** Output Voltage (5V) vs. Temperature ( $I_{LOAD}$  = 10 mA).



**FIGURE 2-14:** GND Current vs. Temperature ( $I_{LOAD} = 10 \text{ mA}$ ).



**FIGURE 2-15:** GND Current vs. Temperature ( $I_{LOAD} = 150 \text{ mA}$ ).

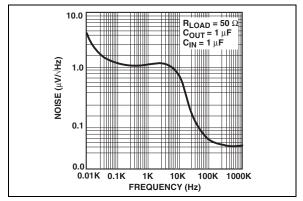


FIGURE 2-16: Output Noise vs. Frequency.

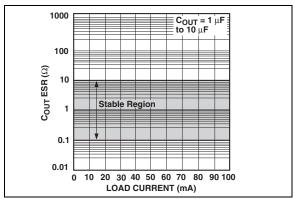
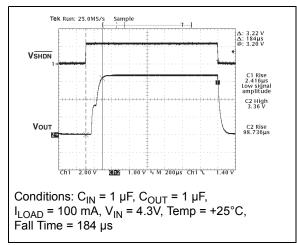
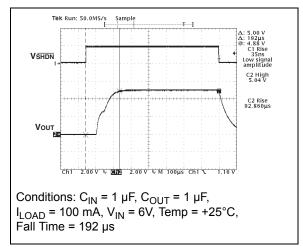


FIGURE 2-17: Stability Region vs. Load Current.

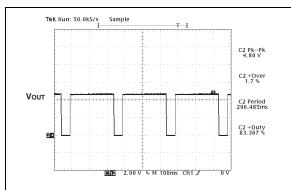


**FIGURE 2-18:** Measure Rise Time of 3.3V LDO.

**Note:** Unless otherwise indicated,  $V_{IN} = V_{OUT} + 1V$ ,  $I_L = 100 \mu A$ ,  $C_L = 3.3 \mu F$ ,  $\overline{SHDN} > V_{IH}$ ,  $T_A = +25^{\circ}C$ .



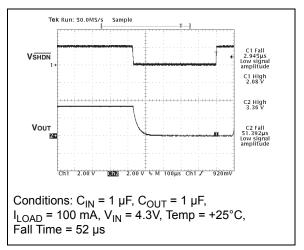
**FIGURE 2-19:** Measure Rise Time of 5.0V LDO.



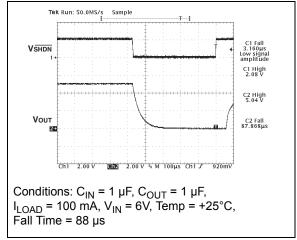
Conditions:  $V_{IN}$  = 6V,  $C_{IN}$  = 0  $\mu$ F,  $C_{OUT}$  = 1  $\mu$ F

 $I_{LOAD}$  was increased until temperature of die reached about 160°C, at which time integrated thermal protection circuitry shuts the regulator off when die temperature exceeds approximately 160°C. The regulator remains off until die temperature drops to approximately 150°C.

**FIGURE 2-20:** Thermal Shutdown Response of 5.0V LDO.



**FIGURE 2-21:** Measure Fall Time of 3.3V LDO.



**FIGURE 2-22:** Measure Fall Time of 5.0V LDO.

#### 3.0 PIN DESCRIPTIONS

The descriptions of the pins are listed in Table 3-1.

TABLE 3-1: PIN FUNCTION TABLE

Pin No.	Symbol	Description	
1	V <sub>IN</sub>	Unregulated supply input	
2	GND	Ground terminal	
3	SHDN	Shutdown control input	
4	ERROR	Out-of-Regulation Flag (Open-drain output)	
5	V <sub>OUT</sub>	Regulated voltage output	

#### 3.1 Unregulated Supply Input (V<sub>IN</sub>)

Connect unregulated input supply to the  $V_{IN}$  pin. If there is a large distance between the input supply and the LDO regulator, some input capacitance is necessary for proper operation. A 1  $\mu F$  capacitor connected from  $V_{IN}$  to ground is recommended for most applications.

#### 3.2 Ground Terminal (GND)

Connect the unregulated input supply ground return to GND. Also connect the negative side of the 1  $\mu$ F typical input decoupling capacitor close to GND and the negative side of the output capacitor  $C_{OUT}$  to GND.

#### 3.3 Shutdown Control Input (SHDN)

The regulator is fully enabled when a logic-high is applied to  $\overline{SHDN}$ . The regulator enters shutdown when a logic-low is applied to  $\overline{SHDN}$ . During shutdown, output voltage falls to zero,  $\overline{ERROR}$  is open-circuited and supply current is reduced to 0.5  $\mu$ A (max).

#### 3.4 Out Of Regulation Flag (ERROR)

 $\overline{\text{ERROR}}$  goes low when  $V_{\text{OUT}}$  is out-of-tolerance by approximately -5%.

#### 3.5 Regulated Voltage Output (V<sub>OUT</sub>)

Connect the output load to  $V_{OUT}$  of the LDO. Also connect the positive side of the LDO output capacitor as close as possible to the  $V_{OUT}$  pin.

#### 4.0 DETAILED DESCRIPTION

The TC1054, TC1055 and TC1186 are precision fixed output voltage regulators (If an adjustable version is desired, please see the TC1070/TC1071/TC1187 data sheet (DS21353)). Unlike bipolar regulators, the TC1054, TC1055 and TC1186 supply current does not increase with load current.

Figure 4-1 shows a typical application circuit, where the regulator is enabled any time the shutdown input (SHDN) is at or above  $V_{IH}$ , and shutdown (disabled) when SHDN is at or below  $V_{IL}$ .  $\overline{SHDN}$  may be controlled by a CMOS logic gate or I/O port of a microcontroller. If the SHDN input is not required, it should be connected directly to the input supply. While in shutdown, supply current decreases to 0.05  $\mu A$  (typical),  $V_{OUT}$  falls to zero volts, and  $\overline{ERROR}$  is opencircuited.

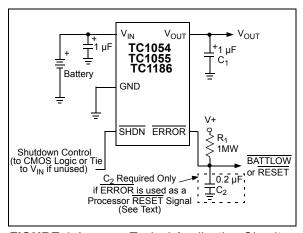


FIGURE 4-1: Typical Application Circuit.

#### 4.1 ERROR Open-Drain Output

ERROR is driven low whenever  $V_{OUT}$  falls out of regulation by more than -5% (typical). This condition may be caused by low input voltage, output current limiting or thermal limiting. The ERROR threshold is 5% below rated  $V_{OUT}$ , regardless of the programmed output voltage value (e.g. ERROR =  $V_{OL}$  at 4.75V (typ.) for a 5.0V regulator and 2.85V (typ.) for a 3.0V regulator). ERROR output operation is shown in Figure 4-2.

Note that  $\overline{\text{ERROR}}$  is active when  $V_{OUT}$  falls to  $V_{TH}$  and inactive when  $V_{OUT}$  rises above  $V_{TH}$  by  $V_{HYS}$ .

As shown in Figure 4-1, ERROR can be used either as a battery low flag or as a processor RESET signal (with the addition of timing capacitor  $C_2$ ).  $R_1 \times C_2$  should be chosen to maintain ERROR below  $V_{IH}$  of the processor RESET input for at least 200 ms to allow time for the system to stabilize. Pull-up resistor  $R_1$  can be tied to  $V_{OUT}$ ,  $V_{IN}$  or any other voltage less than  $(V_{IN} + 0.3V)$ .

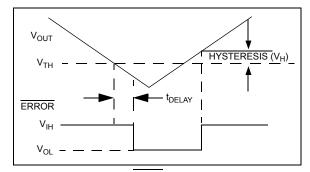


FIGURE 4-2: Error Output Operation.

#### 4.2 Output Capacitor

A  $1 \mu F$  (minimum) capacitor from  $V_{OUT}$  to ground is recommended. The output capacitor should have an effective series resistance greater than  $0.1\Omega$  and less than  $5.0\Omega$ , with a resonant frequency above 1 MHz. A 1 μF capacitor should be connected from V<sub>IN</sub> to GND if there is more than 10 inches of wire between the regulator and the AC filter capacitor or if a battery is used as the power source. Aluminum electrolytic or tantalum capacitor types can be used (Since many aluminum electrolytic capacitors freeze at approximately -30°C, solid tantalums are recommended for applications operating below -25°C.). When operating from sources other than batteries, supply-noise rejection and transient response can be improved by increasing the value of the input and output capacitors and employing passive filtering techniques.

#### 5.0 THERMAL CONSIDERATIONS

#### 5.1 Thermal Shutdown

Integrated thermal protection circuitry shuts the regulator off when die temperature exceeds 160°C. The regulator remains off until the die temperature drops to approximately 150°C.

#### 5.2 Power Dissipation

The amount of power the regulator dissipates is primarily a function of input voltage, output voltage and output current. The following equation is used to calculate worst case actual power dissipation:

#### **EQUATION 5-1:**

$$P_D \approx (V_{INMAX} - V_{OUTMIN})I_{LOADMAX}$$

Where

P<sub>D</sub> = Worst case actual power dissipation

 $V_{INMAX}$  = Maximum voltage on  $V_{IN}$ 

V<sub>OUTMIN</sub> = Minimum regulator output voltage

I<sub>LOADMAX</sub> = Maximum output (load) current

The maximum allowable power dissipation (Equation 5-2) is a function of the maximum ambient temperature ( $T_{AMAX}$ ), the maximum allowable die temperature ( $T_{JMAX}$ ) and the thermal resistance from junction-to-air ( $\theta_{JA}$ ). The 5-Pin SOT-23 package has a  $\theta_{JA}$  of approximately 220°C/Watt.

#### **EQUATION 5-2:**

$$P_{DMAX} = \frac{(T_{JMAX} - T_{AMAX})}{\theta_{JA}}$$

Where all terms are previously defined.

Equation 5-1 can be used in conjunction with Equation 5-2 to ensure regulator thermal operation is within limits.

For example:

Given:

 $V_{INMAX}$  = 3.0V +5%  $V_{OUTMIN}$  = 2.7V - 2.5%

 $I_{LOADMAX}$  = 40 mA  $T_{JMAX}$  = +125°C  $T_{AMAX}$  = +55°C

Find: 1. Actual power dissipation

2. Maximum allowable dissipation

#### Actual power dissipation:

$$P_D \approx (V_{INMAX} - V_{OUTMIN})I_{LOADMAX}$$
  
=  $[(3.0 \times 1.05) - (2.7 \times 0.975)]40 \times 10^{-3}$   
=  $20.7 mW$ 

Maximum allowable power dissipation:

$$\begin{split} P_{DMAX} &= \frac{(T_{JMAX} - T_{AMAX})}{\theta_{JA}} \\ &= \frac{(125 - 55)}{220} \\ &= 318 mW \end{split}$$

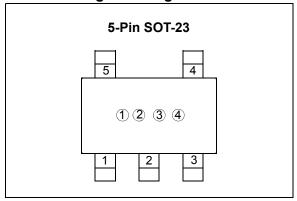
In this example, the TC1054 dissipates a maximum of 20.7 mW; below the allowable limit of 318 mW. In a similar manner, Equation 5-1 and Equation 5-2 can be used to calculate maximum current and/or input voltage limits.

#### 5.3 Layout Considerations

The primary path of heat conduction out of the package is via the package leads. Therefore, layouts having a ground plane, wide traces at the pads and wide power supply bus lines combine to lower  $\theta_{JA}$  and, therefore, increase the maximum allowable power dissipation limit.

#### 6.0 PACKAGING INFORMATION

#### 6.1 Package Marking Information



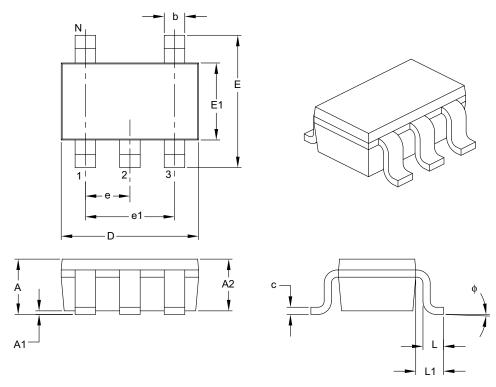
1 & 2 represents part number code + temperature range and voltage

(V)	TC1054 Code	TC1055 Code	TC1186 Code
1.8	CY	DY	PY
2.5	C1	D1	P1
2.6	CT	DT	PV
2.7	C2	D2	P2
2.8	CZ	DZ	PZ
2.85	C8	D8	P8
3.0	C3	D3	P3
3.3	C4	D4	P5
3.6	C9	D9	P9
4.0	C0	D0	P0
5.0	C6	D6	P7

- ③ represents year and quarter code
- ④ represents lot ID number

#### 5-Lead Plastic Small Outline Transistor (CT) [SOT-23]

**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



	Units	MILLIMETERS			
	Dimension Limits		NOM	MAX	
Number of Pins	N	5			
Lead Pitch	е	0.95 BSC			
Outside Lead Pitch	e1	1.90 BSC			
Overall Height	A	0.90	_	1.45	
Molded Package Thickness	A2	0.89	_	1.30	
Standoff	A1	0.00	_	0.15	
Overall Width	E	2.20	_	3.20	
Molded Package Width	E1	1.30	_	1.80	
Overall Length	D	2.70	_	3.10	
Foot Length	L	0.10	_	0.60	
Footprint	L1	0.35	_	0.80	
Foot Angle	ф	0°	-	30°	
Lead Thickness	С	0.08	-	0.26	
Lead Width	b	0.20	_	0.51	

#### Notes:

- 1. Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed 0.127 mm per side.
- 2. Dimensioning and tolerancing per ASME Y14.5M.

BSC: Basic Dimension. Theoretically exact value shown without tolerances.

Microchip Technology Drawing C04-091B

#### **APPENDIX A: REVISION HISTORY**

#### **Revision D (February 2007)**

- Corrected standard output voltages on page 1 and in "Product Identification System".
- Added T<sub>DELAY</sub> parameter in DC Characteristics table in "Electrical Characteristics".
- · Changes to Figure 4-2.
- "Packaging Information": Corrected SOT-23 Packaging Information.

#### Revision C (March 2003)

· Undocumented changes.

#### Revision B (May 2002)

• Undocumented changes.

#### Revision A (March 2002)

· Original Release of this Document.

**NOTES:** 

#### PRODUCT IDENTIFICATION SYSTEM

 $\label{thm:condition} \mbox{To order or obtain information, e.g., on pricing or delivery, refer to the factory or the listed sales of fice.}$ 

PART NO	<u>x,x                                   </u>	Examples:
	T T	a) TC1054-1.8VCT713: 1.8V LDO Regulator
	utput Temperature Package	b) TC1054-2.5VCT713: 2.5V LDO Regulator
Vo	oltage Range	c) TC1054-2.6VCT713: 2.6V LDO Regulator
		d) TC1054-2.7VCT713: 2.7V LDO Regulator
Device:	TC1054: 50 mA LDO with Shutdown & Error output	e) TC1054-2.8VCT713: 2.8V LDO Regulator
Device.	TC1055: 100 mA LDO with Shutdown & Error output	f) TC1054-2.85VCT713: 2.85V LDO Regulator
	TC1186: 150 mA LDO with Shutdown & Error output	g) TC1054-3.0VCT713: 3.0V LDO Regulator
		h) TC1054-3.3VCT713: 3.3V LDO Regulator
Output Voltage *:	1.8 = 1.8V "Standard"	i) TC1054-3.6VCT713: 3.6V LDO Regulator
-	2.5 = 2.5V "Standard"	j) TC1054-4.0VCT713: 4.0V LDO Regulator
	2.6 = 2.6V "Standard" 2.7 = 2.7V "Standard"	k) TC1054-5.0VCT713: 5.0V LDO Regulator
	2.8 = 2.8V "Standard" 2.85 = 2.85V "Standard"	a) TC1055-1.8VCT713: 1.8V LDO Regulator
	3.0 = 3.0V "Standard"	b) TC1055-2.5VCT713: 2.5V LDO Regulator
	3.3 = 3.3V "Standard" 3.6 = 3.6V "Standard"	c) TC1055-2.6VCT713: 2.6V LDO Regulator
	4.0 = 4.0V "Standard"	d) TC1055-2.7VCT713: 2.7V LDO Regulator
	5.0 = 5.0V "Standard"	e) TC1055-2.8VCT713: 2.8V LDO Regulator
	*Contact factory for other output voltage options.	f) TC1055-2.85VCT713: 2.85V LDO Regulator
		g) TC1055-3.0VCT713: 3.0V LDO Regulator
Temperature Range: V = -40°C to +125°C		h) TC1055-3.3VCT713: 3.3V LDO Regulator
		i) TC1055-3.6VCT713: 3.6V LDO Regulator
Package:	CT713 = 5L SOT-23, Tape and Reel	j) TC1055-4.0VCT713: 4.0V LDO Regulator
		☐ k) TC1055-5.0VCT713: 5.0V LDO Regulator
		a) TC1186-1.8VCT713: 1.8V LDO Regulator
		b) TC1186-2.5VCT713: 2.5V LDO Regulator
		c) TC1186-2.6VCT713: 2.6V LDO Regulator
		d) TC1186-2.7VCT713: 2.7V LDO Regulator
		e) TC1186-2.8VCT713: 2.8V LDO Regulator
		f) TC1186-2.85VCT713: 2.85V LDO Regulator
		g) TC1186-3.0VCT713: 3.0V LDO Regulator
		h) TC1186-3.3VCT713: 3.3V LDO Regulator
		i) TC1186-3.6VCT713: 3.6V LDO Regulator
		j) TC1186-4.0VCT713: 4.0V LDO Regulator
		k) TC1186-5.0VCT713: 5.0V LDO Regulator

NOTES:

#### Note the following details of the code protection feature on Microchip devices:

- Microchip products meet the specification contained in their particular Microchip Data Sheet.
- Microchip believes that its family of products is one of the most secure families of its kind on the market today, when used in the intended manner and under normal conditions.
- There are dishonest and possibly illegal methods used to breach the code protection feature. All of these methods, to our
  knowledge, require using the Microchip products in a manner outside the operating specifications contained in Microchip's Data
  Sheets. Most likely, the person doing so is engaged in theft of intellectual property.
- Microchip is willing to work with the customer who is concerned about the integrity of their code.
- Neither Microchip nor any other semiconductor manufacturer can guarantee the security of their code. Code protection does not mean that we are guaranteeing the product as "unbreakable."

Code protection is constantly evolving. We at Microchip are committed to continuously improving the code protection features of our products. Attempts to break Microchip's code protection feature may be a violation of the Digital Millennium Copyright Act. If such acts allow unauthorized access to your software or other copyrighted work, you may have a right to sue for relief under that Act.

Information contained in this publication regarding device applications and the like is provided only for your convenience and may be superseded by updates. It is your responsibility to ensure that your application meets with your specifications. MICROCHIP MAKES NO REPRESENTATIONS OR WARRANTIES OF ANY KIND WHETHER EXPRESS OR IMPLIED, WRITTEN OR ORAL, STATUTORY OR OTHERWISE, RELATED TO THE INFORMATION, INCLUDING BUT NOT LIMITED TO ITS CONDITION, QUALITY, PERFORMANCE, MERCHANTABILITY OR FITNESS FOR PURPOSE. Microchip disclaims all liability arising from this information and its use. Use of Microchip devices in life support and/or safety applications is entirely at the buyer's risk, and the buyer agrees to defend, indemnify and hold harmless Microchip from any and all damages, claims, suits, or expenses resulting from such use. No licenses are conveyed, implicitly or otherwise, under any Microchip intellectual property rights.

#### **Trademarks**

The Microchip name and logo, the Microchip logo, Accuron, dsPIC, KEELOQ, KEELOQ logo, microID, MPLAB, PIC, PICmicro, PICSTART, PRO MATE, PowerSmart, rfPIC, and SmartShunt are registered trademarks of Microchip Technology Incorporated in the U.S.A. and other countries.

AmpLab, FilterLab, Linear Active Thermistor, Migratable Memory, MXDEV, MXLAB, PS logo, SEEVAL, SmartSensor and The Embedded Control Solutions Company are registered trademarks of Microchip Technology Incorporated in the U.S.A.

Analog-for-the-Digital Age, Application Maestro, CodeGuard, dsPICDEM, dsPICDEM.net, dsPICworks, ECAN, ECONOMONITOR, FanSense, FlexROM, fuzzyLAB, In-Circuit Serial Programming, ICSP, ICEPIC, Mindi, MiWi, MPASM, MPLAB Certified logo, MPLIB, MPLINK, PICkit, PICDEM, PICDEM.net, PICLAB, PICtail, PowerCal, PowerInfo, PowerMate, PowerTool, REAL ICE, rfLAB, rfPICDEM, Select Mode, Smart Serial, SmartTel, Total Endurance, UNI/O, WiperLock and ZENA are trademarks of Microchip Technology Incorporated in the U.S.A. and other countries.

 $\ensuremath{\mathsf{SQTP}}$  is a service mark of Microchip Technology Incorporated in the U.S.A.

All other trademarks mentioned herein are property of their respective companies.

@ 2007, Microchip Technology Incorporated, Printed in the U.S.A., All Rights Reserved.

Printed on recycled paper.

QUALITY MANAGEMENT SYSTEM

CERTIFIED BY DNV

ISO/TS 16949:2002

Microchip received ISO/TS-16949:2002 certification for its worldwide headquarters, design and wafer fabrication facilities in Chandler and Tempe, Arizona, Gresham, Oregon and Mountain View, California. The Company's quality system processes and procedures are for its PIC® MCUs and dsPIC® DSCs, KEELOQ® code hopping devices, Serial EEPROMs, microperipherals, nonvolatile memory and analog products. In addition, Microchip's quality system for the design and manufacture of development systems is ISO 9001:2000 certified.



#### WORLDWIDE SALES AND SERVICE

#### **AMERICAS**

**Corporate Office** 

2355 West Chandler Blvd. Chandler, AZ 85224-6199 Tel: 480-792-7200 Fax: 480-792-7277 Technical Support:

http://support.microchip.com

Web Address: www.microchip.com

**Atlanta** 

Duluth, GA Tel: 678-957-9614 Fax: 678-957-1455

**Boston** 

Westborough, MA Tel: 774-760-0087 Fax: 774-760-0088

Chicago Itasca. IL

Tel: 630-285-0071 Fax: 630-285-0075

Dallas

Addison, TX Tel: 972-818-7423 Fax: 972-818-2924

Detroit

Farmington Hills, MI Tel: 248-538-2250 Fax: 248-538-2260

Kokomo

Kokomo, IN Tel: 765-864-8360 Fax: 765-864-8387

Los Angeles

Mission Viejo, CA Tel: 949-462-9523 Fax: 949-462-9608

Santa Clara

Santa Clara, CA Tel: 408-961-6444 Fax: 408-961-6445

Toronto

Mississauga, Ontario, Canada

Tal. OOF

Tel: 905-673-0699 Fax: 905-673-6509

#### ASIA/PACIFIC

**Asia Pacific Office** 

Suites 3707-14, 37th Floor Tower 6, The Gateway Habour City, Kowloon Hong Kong

Tel: 852-2401-1200 Fax: 852-2401-3431

**Australia - Sydney** Tel: 61-2-9868-6733 Fax: 61-2-9868-6755

China - Beijing

Tel: 86-10-8528-2100 Fax: 86-10-8528-2104

**China - Chengdu** Tel: 86-28-8665-5511 Fax: 86-28-8665-7889

China - Fuzhou

Tel: 86-591-8750-3506 Fax: 86-591-8750-3521

China - Hong Kong SAR Tel: 852-2401-1200

Fax: 852-2401-3431 China - Qingdao

Tel: 86-532-8502-7355 Fax: 86-532-8502-7205

China - Shanghai Tel: 86-21-5407-5533

Fax: 86-21-5407-5066

**China - Shenyang** Tel: 86-24-2334-2829 Fax: 86-24-2334-2393

China - Shenzhen

Tel: 86-755-8203-2660 Fax: 86-755-8203-1760

China - Shunde

Tel: 86-757-2839-5507 Fax: 86-757-2839-5571

China - Wuhan

Tel: 86-27-5980-5300 Fax: 86-27-5980-5118

China - Xian

Tel: 86-29-8833-7250 Fax: 86-29-8833-7256

#### ASIA/PACIFIC

India - Bangalore

Tel: 91-80-4182-8400 Fax: 91-80-4182-8422

India - New Delhi

Tel: 91-11-4160-8631 Fax: 91-11-4160-8632

India - Pune

Tel: 91-20-2566-1512 Fax: 91-20-2566-1513

**Japan - Yokohama** Tel: 81-45-471- 6166

Fax: 81-45-471-6122

Korea - Gumi

Tel: 82-54-473-4301 Fax: 82-54-473-4302

Korea - Seoul

Tel: 82-2-554-7200 Fax: 82-2-558-5932 or 82-2-558-5934

Malaysia - Penang Tel: 60-4-646-8870

Fax: 60-4-646-5086

Philippines - Manila

Tel: 63-2-634-9065

Fax: 63-2-634-9069

**Singapore** Tel: 65-6334-8870

Fax: 65-6334-8850

Taiwan - Hsin Chu
Tol: 886 3 573 0536

Tel: 886-3-572-9526 Fax: 886-3-572-6459 Taiwan - Kaohsiung

Tel: 886-7-536-4818 Fax: 886-7-536-4803

**Taiwan - Taipei** Tel: 886-2-2500-6610 Fax: 886-2-2508-0102

Thailand - Bangkok

Tel: 66-2-694-1351 Fax: 66-2-694-1350

#### **EUROPE**

Austria - Wels

Tel: 43-7242-2244-39 Fax: 43-7242-2244-393 Denmark - Copenhagen

Tel: 45-4450-2828 Fax: 45-4485-2829

France - Paris

Tel: 33-1-69-53-63-20 Fax: 33-1-69-30-90-79

Germany - Munich

Tel: 49-89-627-144-0 Fax: 49-89-627-144-44

Italy - Milan

Tel: 39-0331-742611 Fax: 39-0331-466781

Netherlands - Drunen Tel: 31-416-690399

Fax: 31-416-690340 Spain - Madrid

Tel: 34-91-708-08-90 Fax: 34-91-708-08-91 UK - Wokingham

Tel: 44-118-921-5869 Fax: 44-118-921-5820

12/08/06